

# Transmission des mesures d'un réseau de capteurs environnementaux en bande ISM. Une approche semilogicielle

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# Context and Objectives



## ✓ Context

- Need from the French public institution involved in the Earth Science field
  - Automatic Water quality monitoring



## ✓ Current application limitations

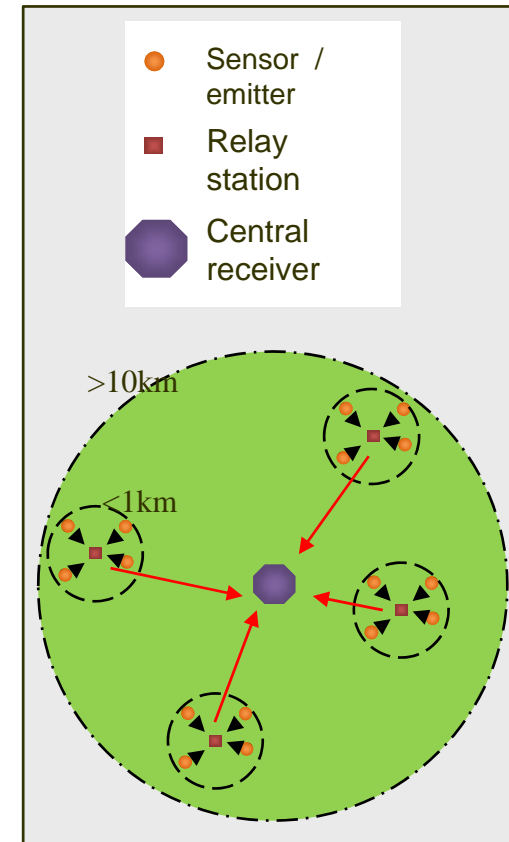
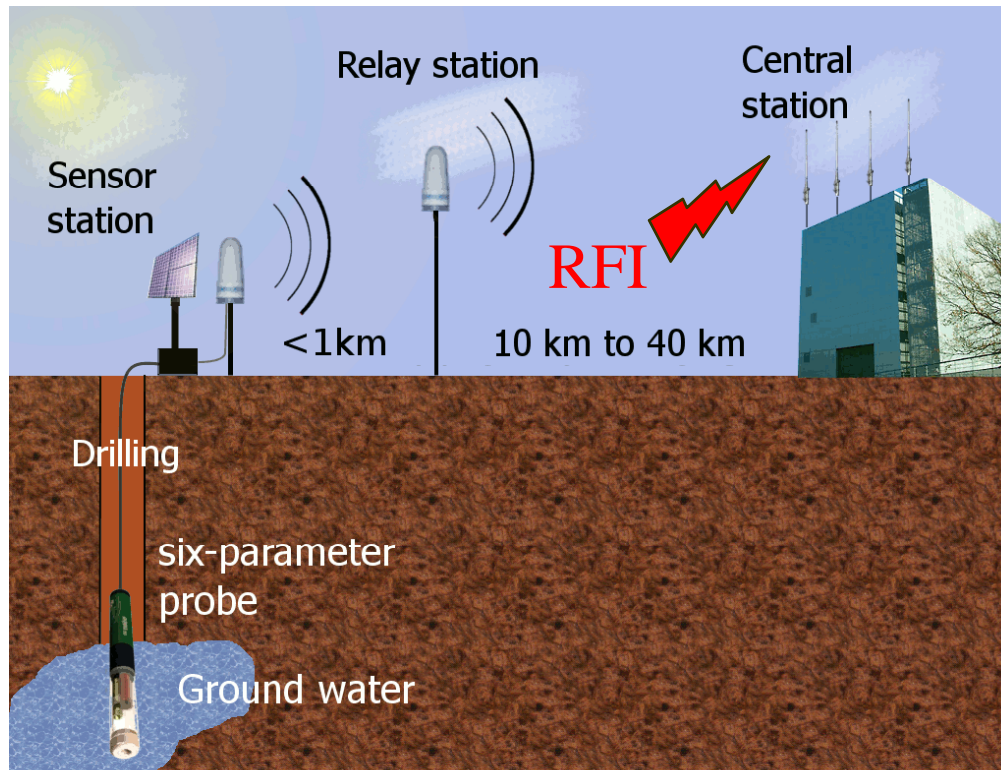
- data loggers requiring manual downloading
- No « real time » sampling (e.g. monthly)



## ✓ Objectives

- Design of a remote sensor network :
  - Autonomous low cost sensors
  - Central receiver in a urban area (radio frequency interference)
  - Emitter/Receiver range >10 km
  - Daily sampling
  - free Band (no fees)

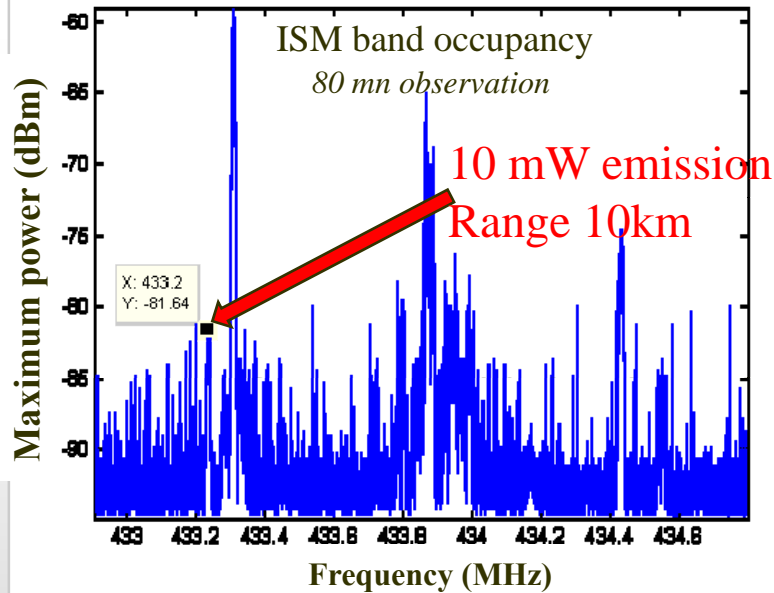
# Sensor network characteristics



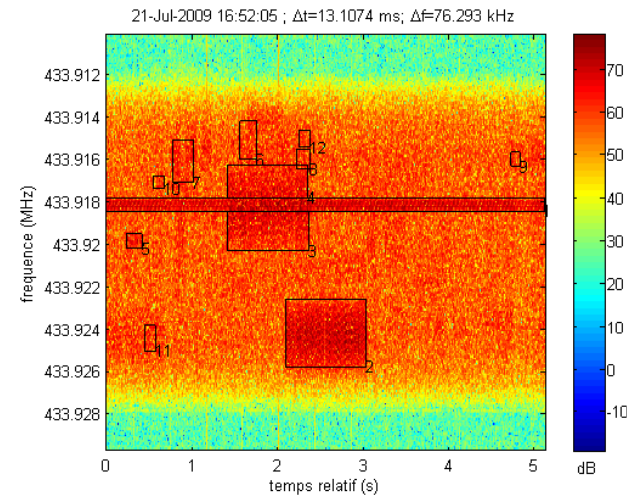
✓ **RF transmission**

- BPSK (ISM band at 433MHz, baud rate=1200 bit/s)
- TDMA

# RFI monitoring in the 433MHz band



## Automatic RFI characterization



n° , when	$f_0$ (MHz)	RFI per hour	$\Delta T$ (s)	$\Delta F$ (kHz)
1, day	433.92	5277	0.3	1.6
2, night	433.92	4549	0.36	1.58
3, day	434.2	2864	0.28	1.31
4, night	434.2	567	0.32	1.33

## ✓ Results

- High RFI occupancy
- 95% RFI < 1s
- Average duration  $\Delta T$ : 30 ms
- Average bandwidth  $\Delta F$ : ~1,3 kHz
- Low activity during the night

# Our approach : selective acquisition and off-line processing



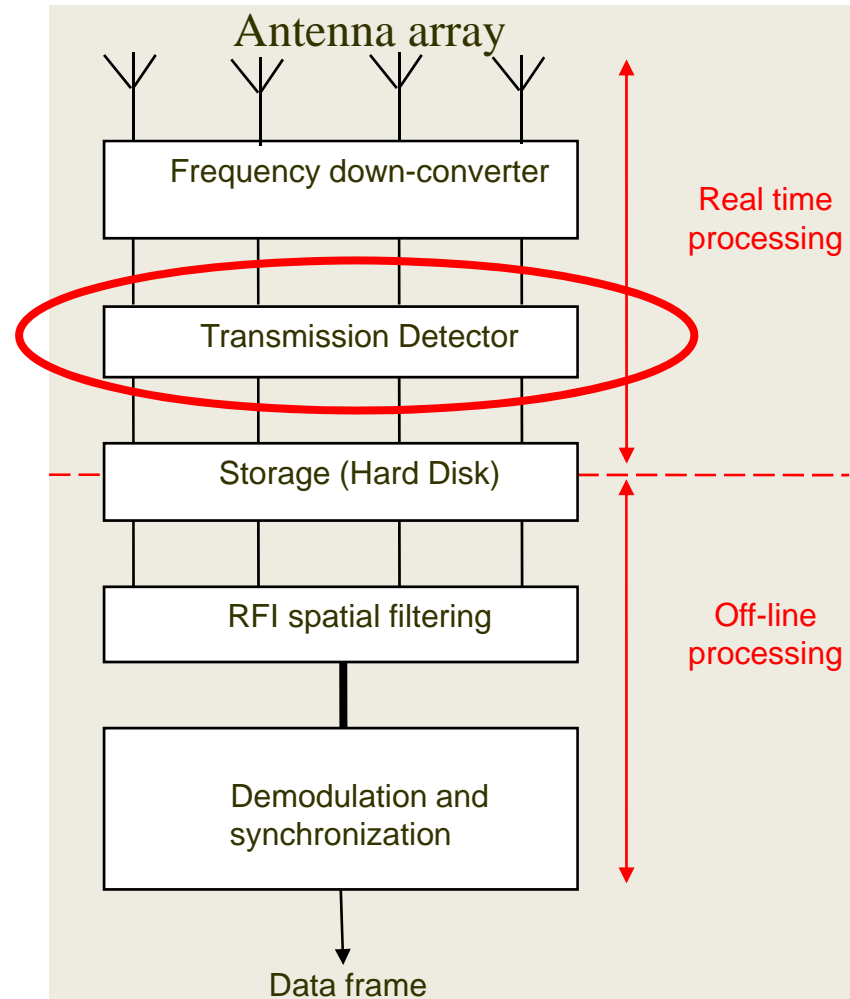
✓ **Specifications**

- Short data frame (<100ms)
- Daily transmission

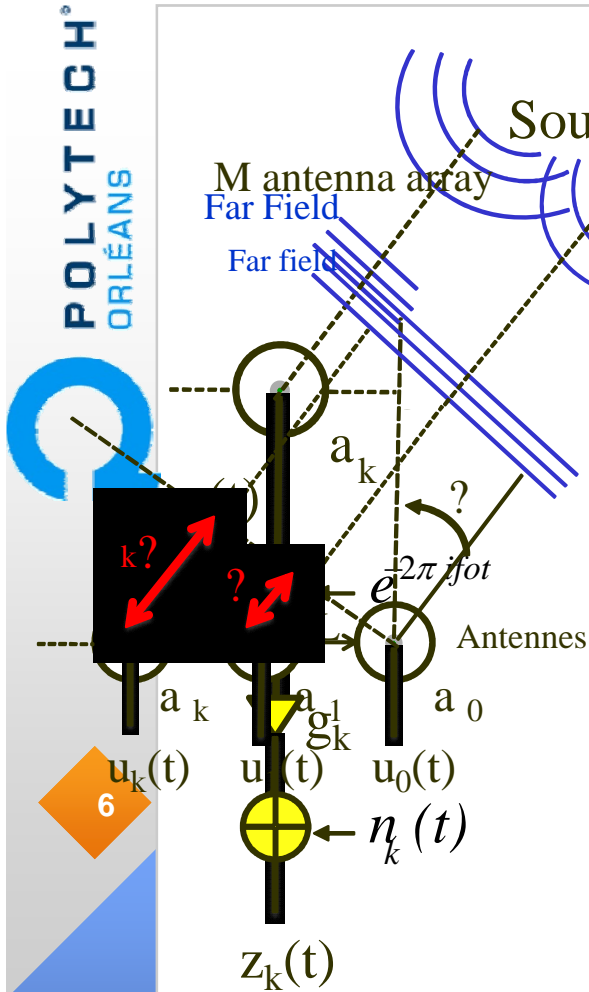
➔ Don't need a continuous reception

✓ **Concept**

- Selective acquisition by detecting sensor transmission
- Data waveform storage
- Antenna array for spatial processing



# Antenna array model



✓

**Signal**

$$s(t)e^{i(2\pi f_0 t + \varphi_0)}$$

$$u_k(t) = s(t - k\tau)e^{i(2\pi f_0(t - k\tau) + \varphi_0)}$$

$$u_k(t) = s(t)e^{-i2\pi f_0 k\tau + i\varphi_0} e^{i2\pi f_0 t}$$

✓

**antenna output**

$$z_k(t) = \underbrace{g_k e^{-i2\pi f_0 k\tau + i\varphi_0}}_{a_{sk}} s(t) + n_k(t)$$

✓

**Antenna array model**

$$\mathbf{z}(t) = \underbrace{\begin{bmatrix} a_{s0} \\ \vdots \\ a_{sM-1} \end{bmatrix}}_{\text{source}} s(t) + \underbrace{\mathbf{a}_r r(t)}_{\text{RFI}} + \underbrace{\mathbf{n}(t)}_{\text{Noise}}$$

## Hypothesis

Far field  
 range > 1km

Narrow band  
 $\frac{\Delta f_0}{f_0} \ll 1$

1 source  $s(t)$   
 1 RFI  $r(t)$

# Cyclostationary detector (1)



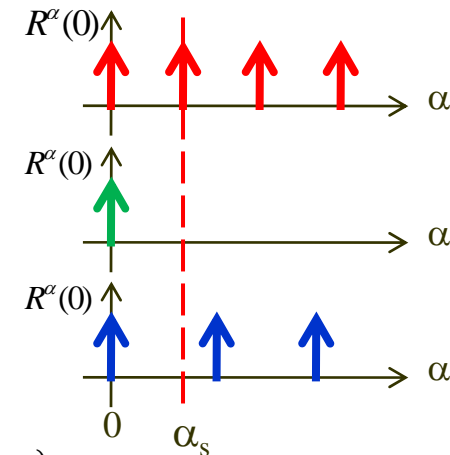
**Cyclic autocorrelation:**  $R_{x,x^*}^\alpha(\tau) = E \left\{ x\left(t + \frac{\tau}{2}\right) x^*\left(t - \frac{\tau}{2}\right) e^{-i2\pi\alpha t} \right\}$

✓ **Different cyclic signatures**

- Source (BPSK)  $\alpha_s = \frac{k}{T_{sym}} \implies R_{s,s^*}^\alpha \neq 0$   
 • Baud rate  $1/T_{sym}$

- Noise = stationary  $\implies R_{n,n^*}^{\alpha_s} = 0$

- RFI  $\alpha_r \neq \alpha_s \implies R_{r,r^*}^{\alpha_s} = 0$



✓ **Multidimensional case**  $\mathbf{R}_{z,z^*}^\alpha = E \left\{ \mathbf{z}(t) \mathbf{z}^H(t) e^{-i2\pi\alpha t} \right\}$

$$\mathbf{R}_{z,z^*}^\alpha = \mathbf{a}_s \underbrace{E \left\{ s s^H e^{-i2\pi\alpha t} \right\}}_{R_{s,s^*}^\alpha} \mathbf{a}_s^H + \mathbf{a}_r \underbrace{E \left\{ r r^H e^{-i2\pi\alpha t} \right\}}_{R_{r,r^*}^\alpha} \mathbf{a}_r^H + E \left\{ \mathbf{n} \mathbf{n}^H e^{-i2\pi\alpha t} \right\}$$

$\alpha = \alpha_s$

$$\mathbf{R}_{z,z^*}^{\alpha_s} = \mathbf{a}_s \mathbf{a}_s^H R_{s,s^*}^{\alpha_s}$$

# Cyclostationary detector (2)



## ✓ Detector criteria

– Singular value decomposition

$$\mathbf{R}_{z,z^*}^\alpha = [r_{ij}] = \mathbf{U}_c \mathbf{\Lambda}_c \mathbf{V}_c^H$$

• With  $\alpha = \alpha_s$ , asymptotically :

$$\mathbf{\Lambda}_c = \begin{bmatrix} \lambda_s & & 0 \\ & 0 & \\ 0 & & 0 \end{bmatrix}_\infty$$

$$\longrightarrow \lambda_{\max} \lesssim \text{threshold}$$

Issue => Real time implementation

– Frobenius norm

$$\|\mathbf{R}_{z,z^*}^\alpha\|_F^2 = \sum_{i=1}^M \sum_{j=1}^M |r_{ij}|^2 = \sum_{k=1}^M |\lambda_k|^2$$

• With  $\alpha = \alpha_s$ , asymptotically :

$$\|\mathbf{R}_{z,z^*}^\alpha\|_F^2 = |\lambda_s|^2$$

$$\longrightarrow \|\mathbf{R}_{z,z^*}^{\alpha_s}\|_F^2 \lesssim \text{threshold}$$



## ✓ Detector performance evaluation

### – Hypothesis

- No transmission:  $H_0 \rightarrow \mathbf{z}(t) = \sqrt{\rho} \mathbf{a}_r r(t) + \sqrt{1-\rho} \mathbf{n}(t)$

- Transmission:  $H_1 \rightarrow \mathbf{z}(t) = \sigma_s \mathbf{a}_s s(t) + \sqrt{\rho} \mathbf{a}_r r(t) + \sqrt{1-\rho} \mathbf{n}(t)$

- 4 detector criteria : 

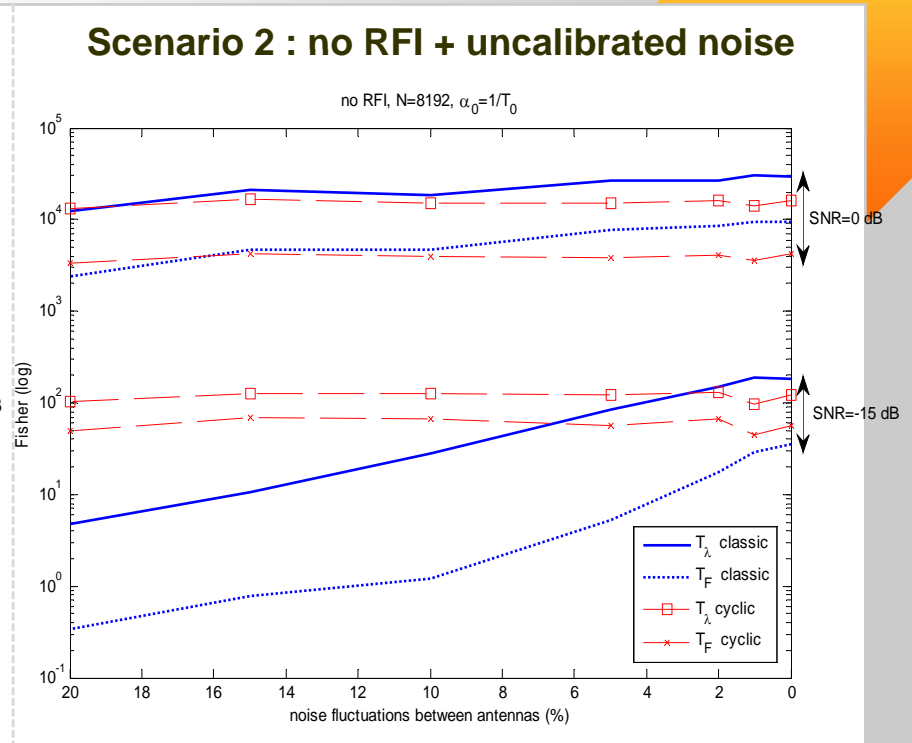
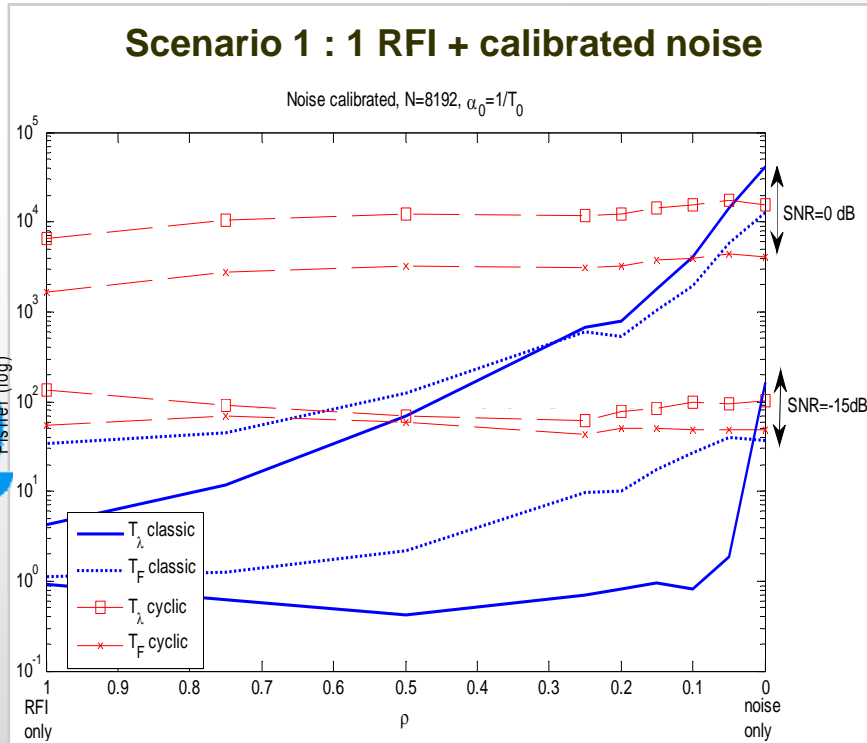
- $SNR = \sigma_s^2$

$T(\mathbf{z})$	$\mathbf{R}_{z,z^*}^0$	$\mathbf{R}_{z,z^*}^{\alpha_s}$
Dominant singular value	$T_\lambda$ classic	$T_\lambda$ cyclic
Frobenius Norm	$T_F$ classic	$T_F$ cyclic

### – Performance criterion : Fisher criteria

$$F = \frac{\left( E_{H_1} [T(\mathbf{z})] - E_{H_0} [T(\mathbf{z})] \right)^2}{\text{Var}_{H_1} [T(\mathbf{z})] + \text{Var}_{H_0} [T(\mathbf{z})]}$$

# Simulations (2)



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## ✓ conclusions

- In practice,  $T_F < T_\lambda$
- Cyclostationary detectors are robust

$$\|\mathbf{R}_{z,z^*}^\alpha\|_F^2 = |\lambda_s|^2 + \underbrace{\sum_{k=1}^M |\lambda_{N,k}|^2}_{\xrightarrow{N \rightarrow \infty} 0}$$

# Filtrage spatial

## ✓ Objectifs

- Focaliser le réseau
- Annuler les brouilleurs

## ✓ Méthode de Capon

- Puissance reçue

$$y(t) = \mathbf{w}^H \mathbf{z}(t)$$

$$P = \left\langle |y(t)|^2 \right\rangle_{\infty} = \mathbf{w}^H \mathbf{R} \mathbf{w}$$

- Critère de Capon

$$\min_{\mathbf{w}} \mathbf{w}^H \mathbf{R} \mathbf{w}$$

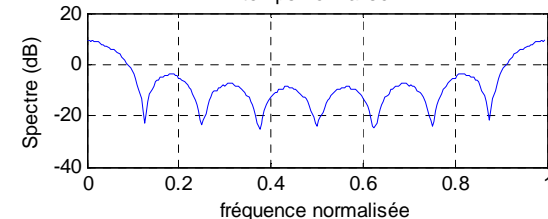
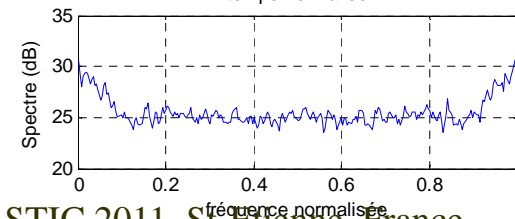
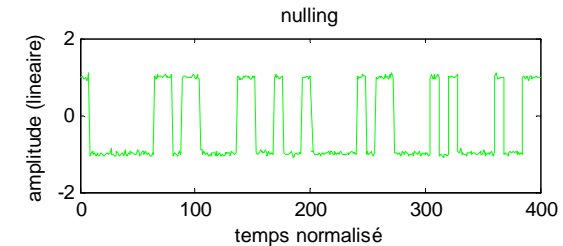
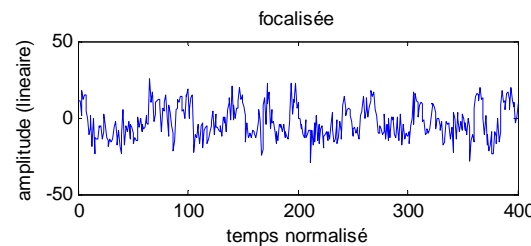
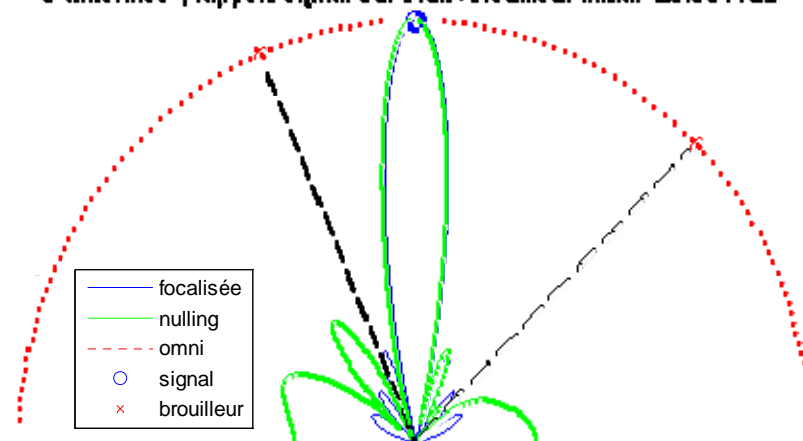
$$\text{avec } \mathbf{w}^H \mathbf{a}_s = 1$$

- Solution

$$\mathbf{w}_o = \frac{\mathbf{R}^{-1} \mathbf{a}_s}{\mathbf{a}_s^H \mathbf{R}^{-1} \mathbf{a}_s}$$

## Simulations

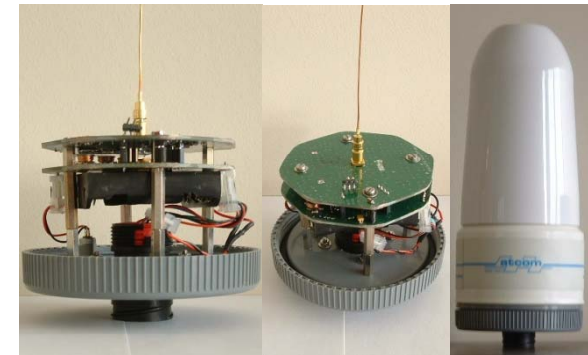
8 antennes , rapport signal sur bruit+brouilleur initial -23.0611dB



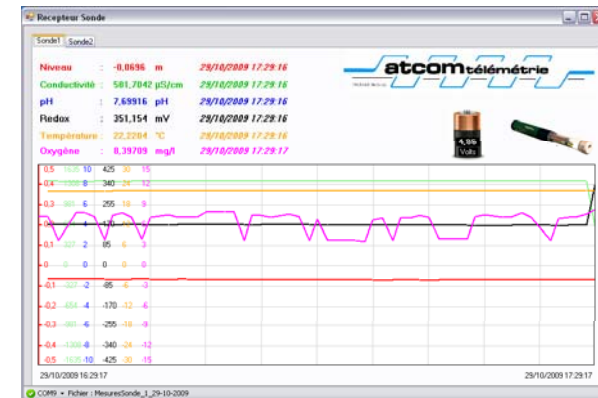
# Conclusions and perspectives (1)



- ✓ Cyclic approach is robust against RFI and uncalibrated array
- ✓ Sensor emitter and relay station have been designed



- ✓ Practical tests with simple receiver (no spatial processing) have shown that long range transmission are very difficult due to RFI
- ✓ Next step : implementation of the proposed approach



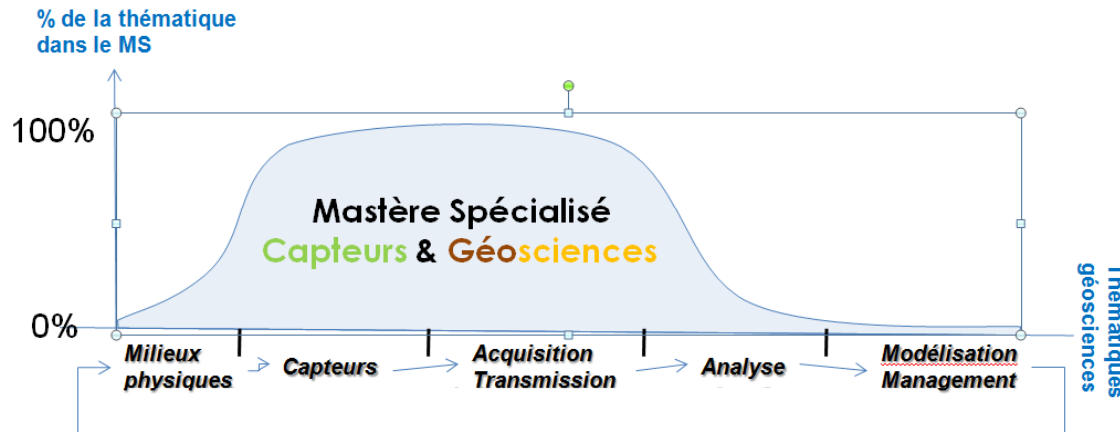
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